

# Characterization of AMOLED pixel circuit without power line

Seon Pyo Hong, Dong Sung Moon & Byung Seong Bae

**Analog Integrated Circuits and Signal Processing**

An International Journal

ISSN 0925-1030

Volume 70

Number 1

Analog Integr Circ Sig Process (2012)

70:157-162

DOI 10.1007/s10470-011-9706-6

**ANALOG INTEGRATED CIRCUITS  
AND SIGNAL PROCESSING**  
An International Journal

Volume 43, Number 1, April 2005

**RFICs**

Thyristor Input-Protection Device Suitable for CMOS RF IC's ..... Jin-Young Choi, Woo Suk Yang, Dongmin Kim and Youngju Kim 5

A Broadband Double-Conversion RF Tuner ..... Karl Stadius, Arto Malinen, Petri Järviö, Petteri Paatsila and Karl Halonen 15

**Data Converters**

1.1 V Low-Power  $\Sigma\Delta$  Modulator for 14-bit, 16 KHz A/D Conversion Using a New Low-Voltage Class-AB Op-amp ..... F. Muñoz, A.P. Vegaleal, R.G. Carvajal, A. Torralba, J. Tombs and J. Ramírez-Angulo 31

**Amplifiers and Filters**

Simplified Modeling of a Multipole Amplifier Using All-Pass Network Functions ..... Yihong Dai, Donald T. Comer, David J. Comer and Darren Korh 39

Design of Square-Root Domain Filters ..... Guo-Jeng Yu, Chun-Yueh Huang, Bin-Da Liu and Jenn-Jiun Chen 49

Fully Differential CMOS Current Feedback Operational Amplifier ..... Soltman A. Mahmoud and Inas A. Awad 61

Single DDCC Biquads with High Input Impedance and Minimum Number of Passive Elements ..... Muhammed A. Ibrahim, Hakan Kuntman and Oguzhan Cicekoglu 71

(continued on back cover)

Available online  
www.springerlink.com

 SPRINGER ISSN: 0925-1030

**Your article is protected by copyright and all rights are held exclusively by Springer Science+Business Media, LLC. This e-offprint is for personal use only and shall not be self-archived in electronic repositories. If you wish to self-archive your work, please use the accepted author's version for posting to your own website or your institution's repository. You may further deposit the accepted author's version on a funder's repository at a funder's request, provided it is not made publicly available until 12 months after publication.**

# Characterization of AMOLED pixel circuit without power line

Seon Pyo Hong · Dong Sung Moon ·  
Byung Seong Bae

Received: 19 September 2010 / Revised: 9 July 2011 / Accepted: 9 July 2011 / Published online: 31 July 2011  
© Springer Science+Business Media, LLC 2011

**Abstract** We fabricated and evaluated the simple active matrix organic light emitting diode (AMOLED) pixel circuits without power line and proved that it is useful for the AMOLED display. Without power line in the pixel circuit we got higher-aperture ratio of emission area than the pixel with power line and the pixel with high aperture ratio can give a long life time due to lower current density of organic light emitting diode. However, the lack of power line requires the verification of the driving scheme of the pixel circuit. After fabrication of two types of AMOLED with and without power line in the pixel, we evaluated the pixel currents under various conditions. The operation of the pixel circuit without power line gave similar characteristics to that of the pixel circuit with power line. By the comparison, we verified that the pixel without power line is acceptable for the application to the AMOLED display combined with feedback compensation scheme for the uniform brightness.

**Keywords** AMOLED · Pixel circuit · OLED · Thin film transistor

## 1 Introduction

The liquid crystal display (LCD) operates by modulating light from back-light and has small viewing angle due to liquid crystal orientation. On the other hand, emissive display does not require back-light and provides wide viewing angle. A promising emissive display is active

matrix organic light emitting diode (AMOLED) which shows wide viewing angle, fast response time, and self emissive characteristics [1].

The brightness of the organic light emitting diode (OLED) is proportional to the amount of current passing through it. Therefore, a pixel of AMOLED is composed of a drive thin film transistor (TFT) as well as a switch TFT to flow a current through the OLED for one frame time. Variation of OLED current results in variation of the brightness of the OLED. Hence, the active matrix OLED pixel array needs to deliver uniform current to the OLED in order to create uniform brightness. Therefore, unlike the LCD, the uniformity of TFT's is quite important for the uniform brightness [2].

To acquire a uniform brightness, a variety of pixel circuits were suggested for AMOLED. There are two types of pixel circuit, one is simple two TFT's and one capacitor (2T1C) type and the other is threshold voltage shift compensation type [3]. The compensation type can reduce the OLED current variation when threshold voltage shifts. Many pixel circuits of compensation type have been reported for AMOLED [4–7].

There are two ways to make OLED pixel structure. One is top emission structure and the other is bottom emission structure [8]. Top emission structure attracts much interest due to its available design margin for increasing fill factor. Top emission needs transparent cathode electrode with low resistance. Bottom emission structure emits light through a glass substrate; therefore, the aperture for light emission is limited by the area of pixel circuit. The area of TFT's, capacitors and signal lines block the light. One way of increasing emitting area is to reduce the design rules. But, it is difficult to reduce the design rule with the existing process equipments. Moreover, smaller line width results in voltage drop across the line, which results in non-uniform brightness [9–11].

S. P. Hong · D. S. Moon · B. S. Bae (✉)  
New IT Engineering College, Hoseo University, Baebang, Asan,  
Chungnam 336-795, Korea  
e-mail: bsbae3@hoseo.edu

Conventional 2T1C circuit is composed of a scan line, a data line, a power line ( $V_{DD}$  line), two TFT's and a capacitor, all of which reduce light emitting aperture in bottom emission structure. Aperture ratio can be increased by elimination of one of them and the pixel circuit without  $V_{DD}$  line was proposed [12]. In the pixel without  $V_{DD}$  line, the power for driving of the OLED is supplied by the scan line. Therefore, we can't use any level of voltages for the scan line because the voltages are the gate voltage of the switch TFT as well as the power voltage of the drive TFT. Therefore, in this paper we evaluated and verified scan voltages to get the appropriate operation for both the switch TFT and drive TFT in the 2T1C type pixel circuit without  $V_{DD}$  line. We fabricated two types of AMOLED pixels with and without power line. After encapsulation with glass to avoid degradation by oxygen or moisture, we evaluated them and proved that the pixel circuit without power line is acceptable for the application to the AMOLED.

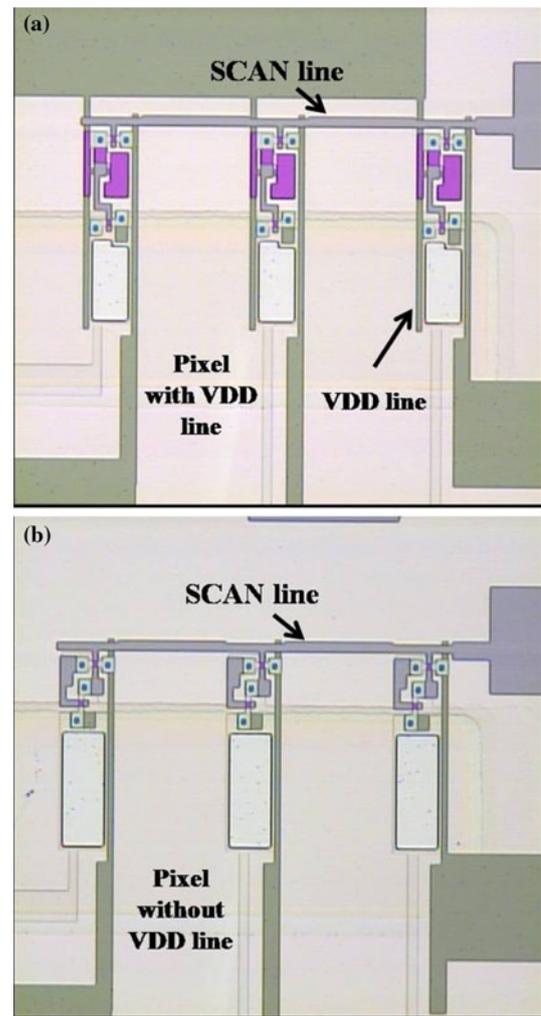
## 2 Experiments

Figure 1(a) and (b) show the microphotographs of the fabricated AMOLED pixels taken from bottom side. Two pixel circuits with and without  $V_{DD}$  line were designed for the low temperature poly silicon TFT process. Figure 1(a) is the conventional 2T1C type and Fig. 1(b) is the 2T1C type circuit without  $V_{DD}$  line. In the 2T1C circuit without  $V_{DD}$  line, scan line was used for supplying  $V_{DD}$  to the drive TFT.

Pixels were fabricated on glass substrates via a low-temperature poly silicon process. Poly silicon was achieved by metal induced lateral crystallization of a 600 Å amorphous silicon thin film by low-pressure chemical vapor deposition (LPCVD). Crystallization was performed at 580°C for 2 h with nickel-catalyst metal. The gate insulator was a 700 Å thickness of  $\text{SiO}_2$  layer deposited by plasma-enhanced chemical vapor deposition (PECVD) with  $\text{SiH}_4$  and  $\text{N}_2\text{O}$  gas at 420°C. The source and drain regions were defined by boron doping.

The gate electrode was Mo metal deposited by sputtering. After deposition of a passivation layer, via hole was formed for contact of indium tin oxide (ITO). ITO was deposited and patterned for the pixel electrode. W/L of the fabricated TFT's of the pixel was 7  $\mu\text{m}$ /7  $\mu\text{m}$ . After the formation of bank with an organic insulator, OLED and cathode were deposited and encapsulation was formed with a glass.

The designed pixel size for both pixels was 65  $\mu\text{m}$   $\times$  195  $\mu\text{m}$ . The widths of the scan line and data line were same as 7  $\mu\text{m}$ . The size of the contact holes for the ITO electrode was 5  $\mu\text{m}$   $\times$  5  $\mu\text{m}$ . The capacitance of the storage capacitor  $C_{st}$  was 0.3 pF. For the pixel with  $V_{DD}$  line, the storage capacitor formed between silicon layer and gate



**Fig. 1** The microphotographs of the fabricated pixel circuits with  $V_{DD}$  line (a); and without  $V_{DD}$  line (b)

layer. Since the photo was taken from bottom side, silicon layer only can be seen. For the pixel without  $V_{DD}$  line, the storage capacitor formed between the ITO and cathode layer. Since the ITO layer is transparent, it is not easy to be seen. The ITO pattern for the storage capacitor was formed at the same step of the pixel ITO, and it is connected to the gate layer through the contact hole. By elimination of  $V_{DD}$  line, fill factor increased as can be seen in Fig. 1(a) and (b). The fill factor of Fig. 1(a) was 22.9% and the fill factor of Fig. 1(b) was 39.7%. The fill factor of the pixel without  $V_{DD}$  line was 16.8% larger than the conventional 2T1C pixel.

Figure 2(a) and (b) show the measurement nodes and input voltage timing diagrams for each pixel circuit. At first, we measured OLED current as changing data voltages. When measuring OLED currents, TFT current meter was disconnected. Next, we measured TFT currents as changing data voltages. OLED was disconnected during the measurement of TFT currents.

The OLED currents depend on the stored voltages in  $C_{st}$  which is the gate voltage of the drive TFT. In the pixel without  $V_{DD}$  line, the  $C_{st}$  was not connected to the scan line to avoid voltage change of the  $C_{st}$  by capacitive coupling to the scan pulse. In the pixel without  $V_{DD}$  line, the scan voltages are applied to both the gate of the switch TFT and the drain of the drive TFT. Scan and data signals are shown in Fig. 1(a) and (b). Scan voltages have two periods, one is scan voltage period and the other is power voltage period. During scan voltage period, the switch TFT turns on and a data voltage is addressed to the gate node of the drive TFT. During the power voltage period, scan voltage becomes off-voltage of the switch TFT. Thus, in the case of the pixel without  $V_{DD}$  line, the off-voltage of the scan line plays a role of the power voltage by which OLED current flows through the drive TFT.

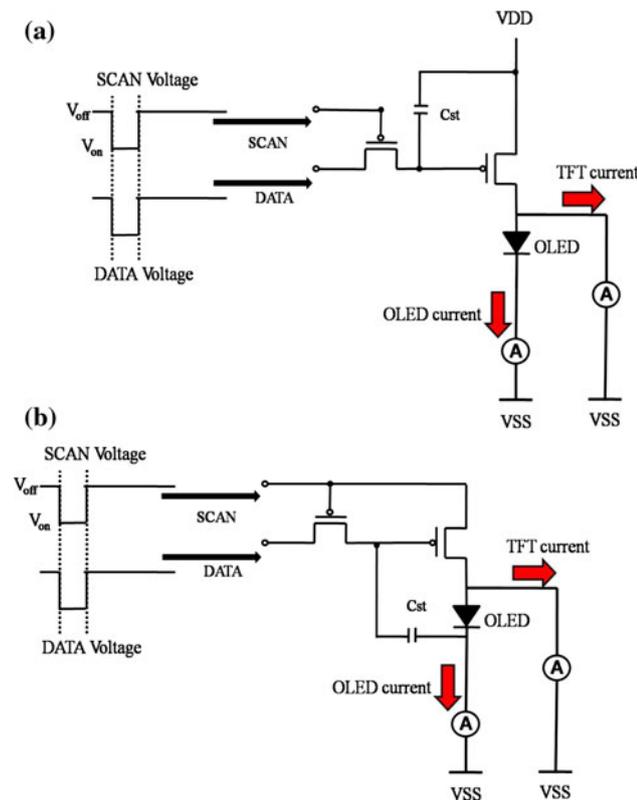
### 3 Results and discussion

Figure 3(a) and (b) show the comparison between OLED currents and TFT currents. The currents were measured for various data voltages from 0 to 5 V. Black squares are OLED currents and white circles are TFT currents.  $V_{DD}$

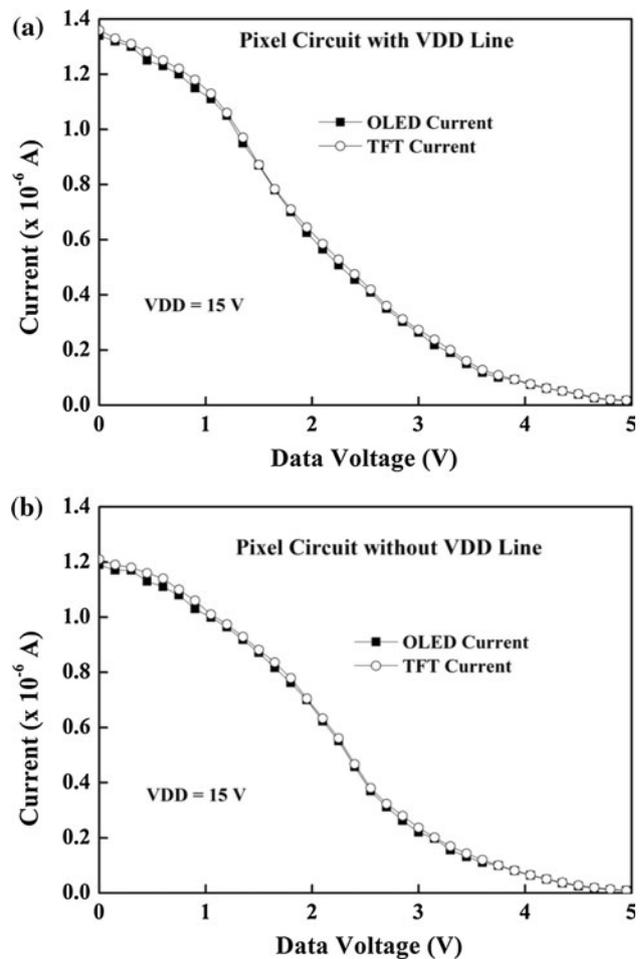
was 15 V and the scan voltages swing from 0 to 15 V. OLED current became zero level around 5 V which was relatively small value because the absolute value of threshold voltage and sub-threshold slope were rather high for the fabricated TFT's. However, the comparison between the pixel with and without  $V_{DD}$  line is possible and it shows that there is not much difference of currents between two pixels with and without  $V_{DD}$  line.

Though OLED currents results in decrease of source-drain voltages of the drive TFT due to the voltage drop across the OLED, the measured OLED currents were almost the same with the TFT currents which is measured without OLED. The same TFT and OLED currents verify that the drive TFT operates in a saturation mode. The pixel without  $V_{DD}$  line also shows the same TFT and OLED currents, which verify that the drive TFT operates in a saturation mode in the pixel without  $V_{DD}$  line.

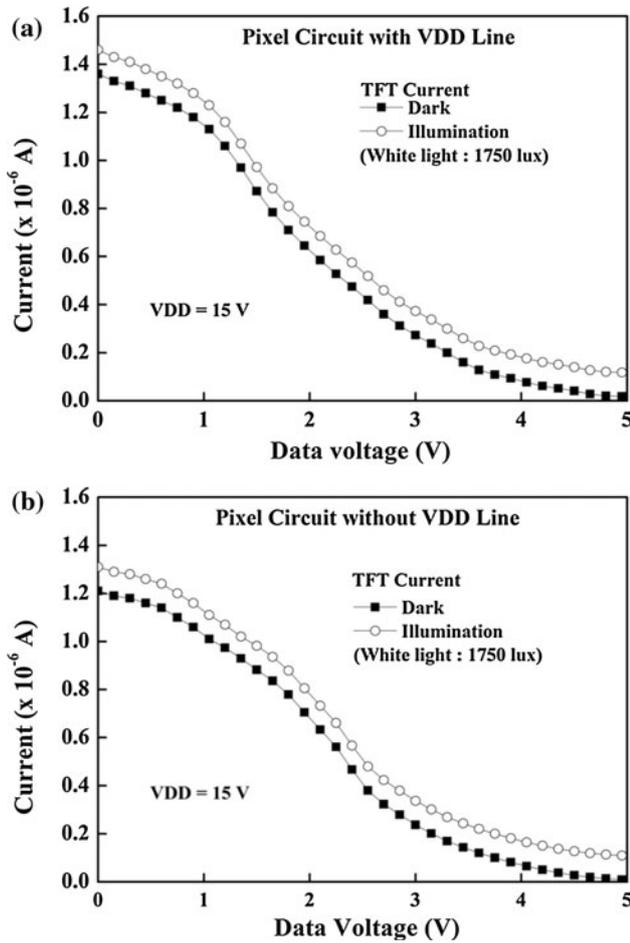
For the lower data voltages below around 1.5 V, the operation regime changed to the linear region because the absolute value of the voltage between the gate and the



**Fig. 2** The measurement nodes and input voltages for the 2T1C pixel with  $V_{DD}$  (a); for the 2T1C pixel without  $V_{DD}$  (b)



**Fig. 3** Comparison between OLED currents and TFT currents for the pixel circuits with  $V_{DD}$  (a); and without  $V_{DD}$  (b)

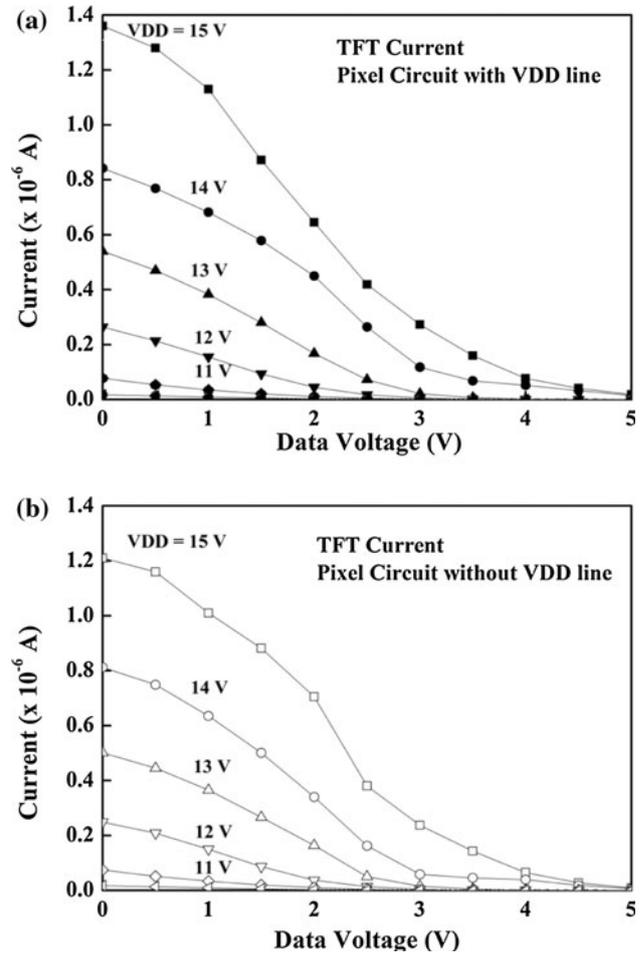


**Fig. 4** Photo and dark TFT currents of both the pixel circuits with  $V_{DD}$  line (a); and without  $V_{DD}$  line (b). Illuminance of the white light was 1750 lx

source/drain increase to the negative sign. As a result, currents do not show quadratic relationship with respect to the data voltages and also the TFT currents and OLED currents become different a little.

We measured the effect of illumination on the pixels. Figure 4(a) and (b) show both the dark and illuminated TFT currents. Figure 4(a) shows TFT currents of the pixel circuit with  $V_{DD}$  line and Fig. 4(b) shows TFT currents of the pixel circuit without  $V_{DD}$  line. Illuminance of white light was 1750 lx. TFT photo currents increase when TFT is illuminated; therefore, the illuminated TFT currents were larger than the dark TFT currents. Since the brightness of OLED is proportional to the amount of current passing through it, display brightness can change by illumination onto the TFT. For uniform brightness it is necessary to prevent illumination onto TFT's. In this measurement, it was verified that the photo-currents for the pixel without  $V_{DD}$  line were similar to those of the pixel with  $V_{DD}$  line.

Figure 5(a) and (b) show TFT currents versus data voltages for various  $V_{DD}$ 's and scan-off voltages.



**Fig. 5** TFT currents dependence on  $V_{DD}$  for both the pixel circuits with  $V_{DD}$  line (a); and without  $V_{DD}$  line (b)

Figure 5(a) shows dependence of the TFT currents on  $V_{DD}$ 's for the pixel circuit with  $V_{DD}$  line. Figure 5(b) shows dependence of the TFT currents on scan-off voltages connected to the drain of the drive. In the case of the pixel circuit without  $V_{DD}$  line, scan-off voltage is supplied to the drive TFT while the switch TFT is off-state.

As the  $V_{DD}$ 's or scan-off voltages decreased, TFT currents also decreased due to the change of the gate-source voltages. In the case of the circuit without  $V_{DD}$  line, scan-off voltages are applied to the switch TFT as well as the drive TFT. Since the off currents of the switch TFT depend on scan-off voltages, we should be careful to set the scan-off voltage. If the leakage currents of the switch TFT of the pixel without  $V_{DD}$  line are large, the  $C_{st}$  node voltage decrease rapidly. Thus, the TFT current should decrease rapidly due to voltage decrease at the  $C_{st}$  node. While varying the scan-off voltages from 15 to 11 V the currents shown in Fig. 5(b) are similar to those of the pixel with  $V_{DD}$  line, which verifies that pixel without  $V_{DD}$  line can be used for the OLED pixel array and allows enough voltage windows of the scan-off voltage. Scan-off voltages resulted

in enough low off-currents of the switch TFT and appropriate drive currents of the drive TFT simultaneously.

We verified that the pixel without  $V_{DD}$  line can be used for the OLED display. However, there are preliminary terms to be concerned with. The first one is the scan driver IC which provides scan voltages and the power voltage in the case of the pixel without  $V_{DD}$  line. Therefore it is necessary to develop the scan driver which has enough current driving capability for the drive TFT's. The second one is the compensation scheme for the unstable TFT characteristics. The 2T1C type can't compensate threshold voltage variation; however, the compensation by video data correction by the sensing the pixel were reported [13] and can be combined with the 2T1C type pixel circuit for the good uniformity of OLED display.

#### 4 Conclusion

We fabricated and analyzed the 2T1C type pixel circuits with and without  $V_{DD}$  line. They were encapsulated with glass after evaporation of OLED. In the pixel without  $V_{DD}$  line, the aperture for the emission area was larger than that of the pixel with  $V_{DD}$  line. For the two types of AMOLED pixels with and without power line, we measured the OLED and TFT currents under the dark and illumination, and TFT currents for various  $V_{DD}$ 's. TFT currents and OLED currents measured under various conditions were similar for both pixels. In the pixel without  $V_{DD}$  line, TFT currents for various data voltages were measured for various scan-off voltages changed from 15 to 11 V. The measured currents were similar to those of the pixel with  $V_{DD}$  line measured with various  $V_{DD}$  voltages changed from 15 to 11 V. The dark and illuminated currents were compared between two pixels and showed similar results, also. In this paper, the operation of the pixel circuits without  $V_{DD}$  line was verified to be used for OLED pixel array with a scan drive IC which can provide high current capability for the drive TFT's. With the compensation by video data correction, the 2T1C type pixel is able to provide uniform brightness against the variation of the TFT characteristics as well as high aperture of light emission area.

**Acknowledgment** This work was supported by grant No. RTI04-01-02 from the Regional Technology Innovation Program of the Ministry of Knowledge Economy (MKE).

#### References

1. Tan, C. W. (1996). An overview of organic electroluminescent material and device. *SID Symposium Digest*, pp. 181–184.
2. Stewart, M., Howell, R. S., Pires, L., & Hatalis, M. K. (2001). Polysilicon TFT technology for active matrix OLED displays. *IEEE Transactions Electron Devices*, 48(5), 845–851.
3. Dawson, R. M. A., Shen, Z., Furst, D. A., Connor, S., Hsu, J., Kane, M. G., Steward, R. G., Ipri, A., King, C. N., Green, P. J., Flegal, R. T., Pearson, S., Barrow, W. A., Dickey, E., Ping, K., Tang, C. W., Van Slyke, S., Chen, F., Shi, J., Sturm, J. C., Lu, M. H. (1998). Design of an improved pixel for a polysilicon active-matrix organic LED display. *SID Symposium Digest*, 1998, pp. 11–15.
4. Jung, S. H., Nam, W. J., & Han, M. K. (2004). A new voltage-modulated AMOLED pixel design compensating for threshold voltage variation in Poly-Si TFT's. *IEEE Electron Device Letters*, 25(25), 690–692.
5. Choi, S. M., & Kwon, O. K. (2003). A self-compensated voltage programming pixel structure for active-matrix organic light emitting diodes. *Proceedings of International Symposium IDW*, 2003, 535–538.
6. Ashtiani, S. J., Chaji, R., & Nahan, A. (2007). AMOLED pixel circuit with electronic compensation of luminance degradation. *Journal of Display Technology*, 3(1), 36–39.
7. Papadopoulos, N. P., Hatzopoulos, A. A., & Papakostas, D. K. (2009). An improved optical feedback pixel circuit. *IEEE Transactions on Electronic Devices*, 56(2), 229–235.
8. Ishibashi, T., Yamada, J., Hirano, T., Iwase, Y., Sato, Y., Nakagawa, R., et al. (2006). Active matrix organic light emitting diode display based on super top emission technology. *Japanese Journal of Applied Physics*, 45(5B), 4392–4395.
9. Jung, M. H., Kim, O., & Chung, H. J. (2006). Voltage distribution of power source in large AMOLED displays. *Journal of the Korean Physical Society*, 48, S5–S9.
10. Lu, H. Y., Chang, T. C., Tai, Y. H., Liu, P. T., & Chi, S. (2007). A new pixel circuit compensating for brightness variation in large size and high resolution AMOLED displays. *Journal of Display Technology*, 3(4), 398–403.
11. Lu, M. M., Hack, M., Hewitt, R., Weaver, M. S., & Brown, J. J. (2008). Power consumption and temperature increase in large area active-matrix OLED displays. *Journal of Display Technology*, 4(1), 47–53.
12. Nam, W. J., Lee, J. H., Kim, C. Y., Lee, H. J., & Han, M. K. (2006). High aperture pixel design employing VDD line elimination for active matrix organic light emitting diode display. *Japanese Journal of Applied Physics*, 45(4), 2433–2436.
13. Min, U. G., & Kwon, O. K. (2009). A video data correction method for the non-uniform electro-optical characteristics of the pixels in AMOLED displays. *Journal of Information Display*, 10(2), 80.



**Seon Pyu Hong** was born in Korea, in 1982. He received the BS and MS degree from Display Engineering of the Hoseo University, Asan, Korea, in 2007 and 2009, respectively. He joined Orion OLED in 2009. And now he is working in Sichun CCO Display Technology Co., LTD in China. His current research interest is organic light emitting diode.



**Dong Sung Moon** received the BS degree in electronic engineering in Pu-kyong National University, Pusan, Korea, in 1991 and the MS degree in Automatic Control engineering in Dong-A University, Pusan, Korea, in 1993. Between 1991 and 1999, he worked at the Daewoo Electronics on development of CRT Semiautomatic system. Between 1999 and 2010, he worked as a senior researcher at the Orion OLED. And now he is Ph.D. student of

Hoseo University and CTO in Sichun CCO Display Technology CO., LTD in China.

and developed micro-display for projection display at ILJIN Display. Since 2006, he is a Professor, New IT Engineering College of the Hoseo University, Asan, Korea.



**Byung Seong Bae** received the BS degree in atomic nuclear engineering from the Seoul National University, Seoul, Korea, in 1984 and the MS and PhD. degrees in applied physics from the Korea Advanced Institute of Science and Technology, Seoul, Korea in 1986, and 1991, respectively. Between 1991 and 1998, he worked at the Samsung Electronics on the development of amorphous and poly-silicon TFT LCD with integrated driver. From 1999 to

2003, he set up the high-temperature poly-silicon TFT LCD factory